



Deep Space Network

211 Wideband Very-Long Baseline Interferometry

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A	4/15/2011	Kurt Liewer	Section 2, 2.1, 2.3 Figure 2, 2.3.3, 2.3.5, and 2.4	<ul style="list-style-type: none"> • Removed references to Mark III VLBI DAT • Replaced all instances of “radio telescope” to “antenna” • Added text noting new DAT system will be replacing old in 2013. • Removed EAC references • Noted termination of Mark-IV tape support • Updated Figure 2 to reflect changes in the system
B	02/09/2015	A. Bedrossian J. S. Border D. Shin	Section 2 and Appendix	<ul style="list-style-type: none"> • Removed DTE, DAT and PCFS • Added DVP and Mark 5C • Added DSS—26, -35, -54 • Updated Figures, Tables, and Appendix
C	02/10/17	J. S. Border D. Shin	Section 2.3	Added DSS-36. Removed DSS-45.

Contents

<u>Paragraph</u>	<u>Page</u>
1 Introduction.....	5
1.1 Purpose.....	5
1.2 Scope.....	5
2 General Information.....	5
2.1 Description of Very-Long Baseline Interferometry.....	5
2.2 Bandwidth Synthesis.....	7
2.3 DSN VLBI Support Equipment.....	8
2.3.1 Signal Reception.....	9
2.3.2 Subreflector Position.....	10
2.3.3 Data Acquisition.....	10
2.3.4 Phase Calibration Tones.....	11
2.3.5 Correlation.....	12
2.4 Experiment Design.....	12
Appendix A.....	14
References.....	14

Illustrations

<u>Figure</u>	<u>Page</u>
Figure 1. Very-Long Baseline Interferometry.....	6
Figure 2. DSN Equipment for Wideband VLBI Support.....	9

Tables

<u>Table</u>	<u>Page</u>
Table 1. Frequency and Polarization Support Capabilities for VLBI.....	10
Table 2. Typical Channels (S and X Bands) for Wideband VLBI.....	13

1 Introduction

1.1 Purpose

This module provides an introduction to the Very-Long Baseline Interferometry (VLBI) astrometric measurement technique and identifies the capabilities and performance of Deep Space Network (DSN) equipment used to support wideband VLBI measurements.

1.2 Scope

The content of this module is limited to a discussion of the data gathering and data processing equipment used by the DSN to support VLBI. Experiment design and the quality of the results are influenced by many other factors that are discussed elsewhere in this handbook. Among these factors are antenna performance parameters discussed in modules 101, 103 and 104, and frequency reference performance discussed in module 304. VLBI experimenters may also be interested in station locations discussed in module 301 and extra-galactic radio source positions provided by module 107. It should be noted that the equipment discussed in this module is the major contributor to the information content of modules 301 and 107.

2 General Information

The DSN uses its VLBI capability to measure the earth orientation parameters that establish the relationship of the Terrestrial Reference Frame containing its stations to the Celestial Reference Frame used for spacecraft navigation. VLBI is also used to determine the relative locations of its stations and the locations of extra-galactic radio sources (EGRSs). The EGRSs are used as reference points for delta-differential one-way ranging (Δ DOR) spacecraft navigation (see module 210). Because the DSN uses equipment that is compatible with that used by the international VLBI community, other agencies can coordinate their observation schedules with DSN scheduled activities and the resultant data can be shared between the DSN and external experimenters.

2.1 Description of Very-Long Baseline Interferometry

Very-Long Baseline Interferometry is a technique for measuring the precise time difference between the arrivals of waveforms originating from an EGRS at two (or more) stations. Figure 1 provides an illustration of the technique. The stations simultaneously observe an EGRS. As they are widely separated, their baseline, B , passes through the Earth. At each station, the instantaneous voltage of the received signal (a random Gaussian process) is recorded in each of several channels; this is known as VLBI data acquisition. The measured difference in arrival time can be converted to an approximate path length difference by multiplying by c , the speed of electromagnetic waves in a vacuum. In actuality, the conversion is much more complex

as the wave front must pass through several layers of the Earth's atmosphere – each of which has a different propagation speed.

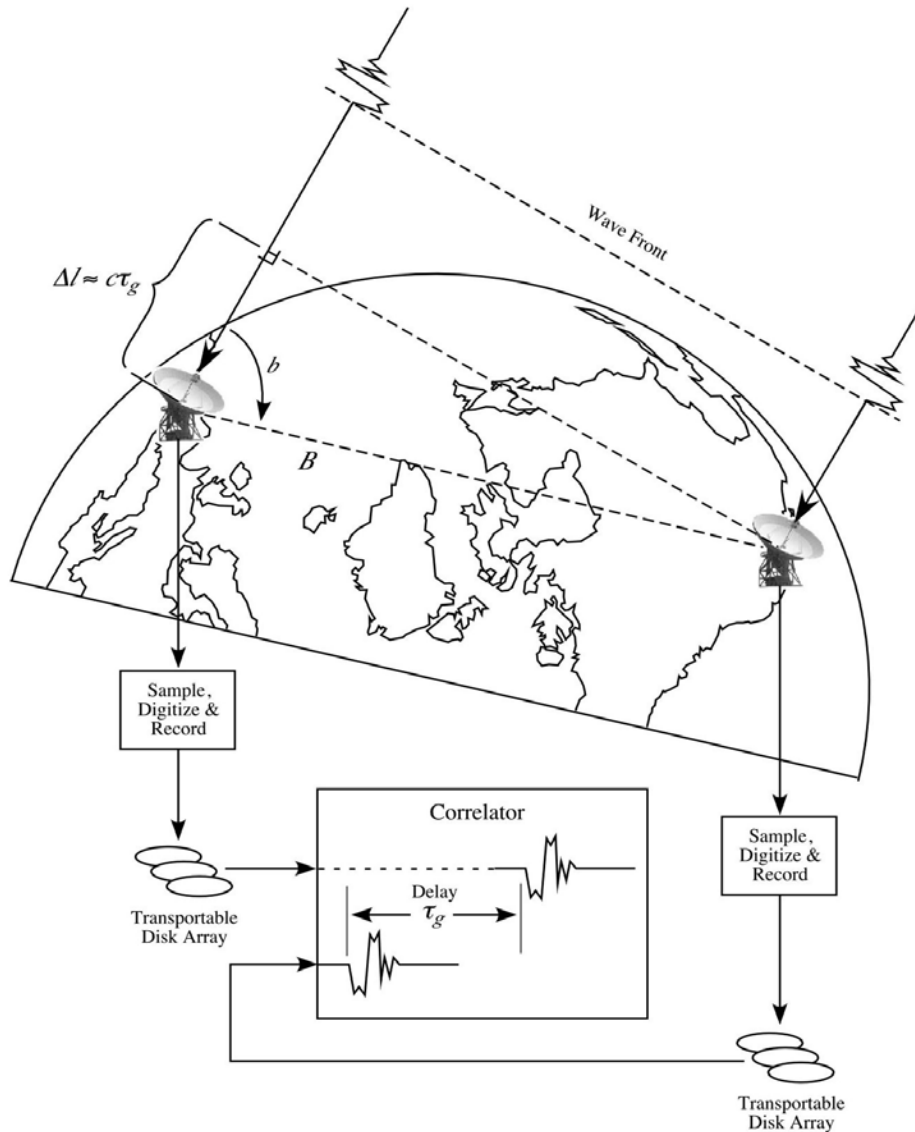


Figure 1. Very-Long Baseline Interferometry

The recorded data from the participating antennas are brought together at a correlator. The data from matching channels at the different radio telescopes are cross-correlated to determine the geometric time delay, τ_g ; this is known as VLBI data correlation. After converting the time delay to a path length by a process that takes into account atmospheric delay and other effects, the angular relationship between the baseline of the two stations and the radio source can be determined by the relationship $b = \cos^{-1}(\Delta l/B)$. It is important to note that this is not a complete solution for the angular position in the sky of the radio source. This calculation only provides information about the location of the radio source in the plane defined by the

interferometer baseline and the direction of arrival. The correlation of data from the observation of additional radio sources is necessary to completely define the baseline. By making sufficient observations and assuming some reference, it is possible to calculate a singular solution that estimates each station's coordinates, the angular positions of all observed sources, the offset and rate differences between the two station frequency references, and several parameters relating to propagation. Additional observations above the minimum enable statistical uncertainties in the estimation to be reduced. A complete discussion of the process is contained in Reference 1.

The precision of a VLBI measurement improves with the length of the baseline. The location of the DSN stations provides baselines of 66% (Goldstone to Madrid) and 83% (Goldstone to Canberra) of the Earth's diameter. The third baseline, Canberra to Madrid, is 98% of the Earth's diameter but provides very limited mutual visibility (see module 301). To mitigate mutual visibility effects, additional antennas are often involved in VLBI measurements allowing a station between DSN locations to serve as a "relay" to tie-together observations made in different portions of the sky during an observation session.

2.2 *Bandwidth Synthesis*

The signal-to-noise ratio obtained by correlating the signals from two observing stations depends on the characteristics of the stations, the bandwidth of the channel being correlated and the observation time. It can be calculated from the following expression.

$$SNR_{CH} = 2.05 \times 10^{-4} (\gamma_{\nu} S) D_1 D_2 \sqrt{\frac{\eta_1 \eta_2 B_{CH} T}{T_{Sys1} T_{Sys2}}} \quad (1)$$

where

- $(\gamma_{\nu} S)$ = correlated flux density where S is the total flux density (jansky)
- D_1, D_2 = antenna diameters (m)
- η_1, η_2 = antenna efficiencies
- B_{CH} = channel bandwidth (Hz)
- T = observation time (s)
- T_{Sys1}, T_{Sys2} = antenna system noise temperatures (K)

The factor 2.05×10^{-4} includes all normalization constants and allowances for degradation due to one-bit quantization and the fact that only one polarization is received.

When multiple channels are being correlated simultaneously and they have approximately the same SNR_{CH} , the post-correlation SNR for use in equation (3) can be calculated by

$$SNR = \sqrt{n} \cdot SNR_{CH} \quad (2)$$

where n is the number of channels in the correlation.

The error in delay measurement resulting from the correlation of data from two VLBI stations is expressed by

$$\sigma = \frac{c}{2\pi \cdot SNR \cdot BW} \quad (3)$$

where

c = the speed of electromagnetic propagation

SNR = the post-correlation signal-to-noise ratio

BW = the observation bandwidth

It can be seen from equation (3) that an extremely large data volume for transport to the common correlation site would be needed if the entire bandwidth were recorded. For example, if a post-correlation SNR of 40 (16 dB) is available, a bandwidth of 125 MHz would be required to achieve a delay error of 1 cm.

The data volume is significantly reduced by the technique of *bandwidth synthesis* (Reference 2). This technique uses several narrow channels spread across the observation bandwidth to achieve an *RMS bandwidth* of

$$BW_{RMS} = \sqrt{\frac{\sum_{CH=1}^n (f_{CH} - f_{AVG})^2}{n}} \quad (4)$$

where

f_{CH} = center frequency of each channel used in the bandwidth synthesis

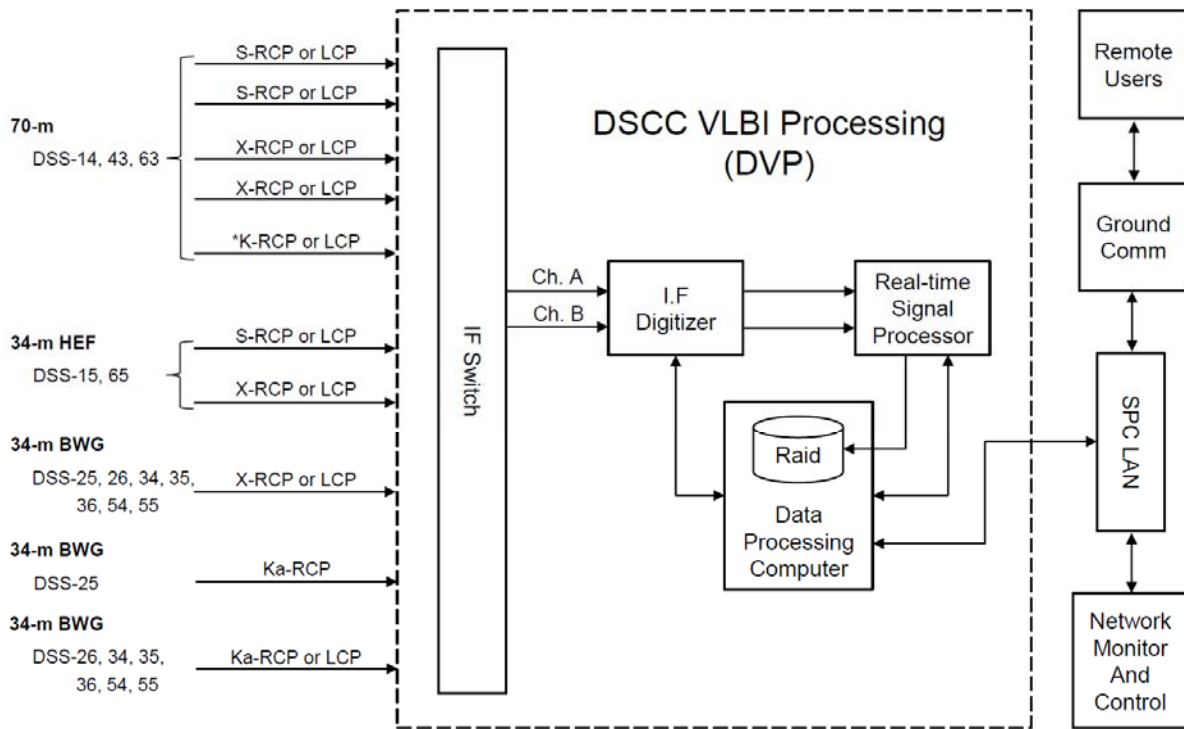
f_{AVG} = the average of all channel center frequencies used in the bandwidth synthesis

n = the number of channels

As an example of the efficiency gained by bandwidth synthesis, a typical X-band VLBI measurement will use the eight 2-MHz channel pairs shown in Table 2 corresponding to an RMS bandwidth of 140 MHz. By employing hard limiting, one-bit quantization, and a sampling rate equal to twice the bandwidth of each channel, the total recorded bandwidth is reduced to 64 MHz.

2.3 DSN VLBI Support Equipment

A diagram showing the major items of DSN equipment used to support wideband VLBI is provided by Figure 2. All DSN 34-m HEF antennas and 70-m antennas are available for VLBI data acquisition. In addition, some 34-m BWG antennas are used for VLBI.



- Note * DSS-63 only
1. S and X-band PCG tones are available only at 70-m and 34-m HEF sites
 2. X and Ka-band PCG tones are available only at DSS-25

Figure 2. DSN Equipment for Wideband VLBI Support

Received signals are processed by the Deep Space Communication Complex (DSCC) VLBI Processing (DVP) assembly. The VLBI data is recorded on a Mark 5 disk array. The DVP interfaces directly with the DSN Network Monitor and Control Subsystem.

2.3.1 Signal Reception

Signals that are collected by the antenna are amplified by a cryogenically-cooled low-noise amplifier (LNA). For every observation band, except L, the LNA is followed by a downconverter that translates the radio-frequency signal to an intermediate frequency (IF) to be forwarded to the Signal Processing Center (SPC). The downconverters are located on the antenna and are derived from the station frequency standard. Their frequencies are 2.0 GHz for S band, 8.1 GHz for X band, and 31.7 GHz for Ka band. The L-band signal at a 70-m antenna is upconverted to S-band using a 620-MHz local oscillator and then switched into one of the S-band downconverters. From that point forward, it follows the path normally taken by a signal originating in the S band.

DSS-25 and -26 at the Goldstone DSCC, DSS-34, -35, and -36 at the Canberra DSCC, and DSS-54 and -55 at the Madrid DSCC are available for Ka-band VLBI support. However, the Ka-band configuration of DSS-25 is somewhat different. The Ka-band feed of DSS-25 supports RCP but does not support LCP. The frequency and polarization capabilities of the DSN for VLBI are summarized in Table 1.

Table 1. Frequency and Polarization Support Capabilities for VLBI

Subnet or Station	S-band	X-band	K-band*	Ka-band
70-m subnet (DSS-14, 43, & 63)	RCP and LCP (2200–2300 MHz)	RCP and LCP (8200–8600 MHz)	RCP or LCP (17-27 GHz)	–
34-m HEF subnet (DSS-15 and 65)	RCP or LCP (2200–2300 MHz)	RCP or LCP (8200–8600 MHz)		–
34-m BWG (DSS-25)	–	RCP or LCP (8200–8600 MHz)		RCP (31.8–32.3 GHz)
34-m BWG (DSS-26, 34, 35**, 36, 54, & 55)	–	RCP or LCP (8200–8600 MHz)		RCP or LCP (31.8–32.3 GHz)

* DSS-63 only

** DSS-35 X-band bandwidth is limited to 8225-8600 MHz

2.3.2 *Subreflector Position*

Gravity deformation of the primary reflector and the structure that supports the subreflector as the antenna moves in elevation causes a change in the antenna focal position. To compensate for this, the subreflector undergoes a programmed motion relative to the feed to achieve the maximum antenna gain at every elevation angle. However, this motion introduces a time-varying phase delay to the received signal that is undesirable for wideband VLBI.

During wideband VLBI data acquisition at the 34-m HEF and 70-m antennas, the subreflector motion is usually disabled and the subreflector placed in a position that gives the best gain for a 45° elevation angle. This eliminates the time-varying phase delay associated with subreflector motion. However, it also causes some loss of antenna gain when the antenna is not at 45°.

Subreflector motion should not be disabled when a BWG antenna is used for wideband VLBI observations in the Ka band. This is because gravity deformation also affects antenna pointing by an amount that is significant at Ka-band although not so at S- or X-bands and subreflector motion must be enabled in order to have the antenna point as commanded. The resultant phase delay changes can be modeled and removed in post processing

2.3.3 *Data Acquisition*

The DVP provides for digitization, downconversion, filtering, and recording of the data. The analog IF signals from each antenna are connected to the IF Switch Input of the

DVP. The DVP then selects any two of them as inputs, and filters each signal to a bandwidth of 100 to 600 MHz. Each input is digitized at 1280 Msamples/sec with 8-bit resolution.

Either or both of these two digital IFs can be routed to the Realtime Signal Processor. The Realtime Signal Processor can output up to 16 baseband channels. Each baseband channel is obtained by digitally downconverting and filtering a selected input to a specified center frequency and bandwidth. Complex (I/Q) samples are initially obtained for each baseband channel. If specified, the I/Q samples can be converted to real upper sideband/lower sideband (USB/LSB) samples.

The outputs of the baseband channel are placed into recording frames along with timing and parity data, and sent to the Mark 5C [Reference 5] disk arrays. Supported channel bandwidths include 2, 4, 8, 16, and 32 MHz. The maximum supported sample rate is currently 2048 Msamples/sec. For example, 16 complex channels with 32 MHz bandwidth and 2-bit sample resolution could be recorded. Bit configurations of 4, 8 and 16 are also supported, as long as: $\text{num_channels} \times \text{num_bits} \times 2 = 64$, and the maximum supported bitrate is not exceeded. For example, 4 complex channels with 8 bits is supported with any bandwidth up to 32 MHz. Likewise, 2 complex channels with 16-bits is supported with any bandwidth up to 32 MHz. Also, all channels must be configured identically. Note that 1-bit recordings are not currently supported.

2.3.4 Phase Calibration Tones

The 34-m HEF and 70-m antennas include S-band and X-band phase calibration generators (PCGs) for all of their supported frequency bands (Reference 4). The PCGs inject a set of tones, uniformly spaced in frequency, into the receive signal path ahead of the LNA. The purpose of these tones is to permit calibration of irregularities in phase delay occurring along the signal path to the DVP. The tones are spaced 1.0 MHz apart and are derived from the station frequency reference. The power in the phase calibration tones is less than 1% of the signal power in any given bandwidth. With such a small power, the phase calibration tones do not interfere with the VLBI measurements. The 1.0 MHz spacing guarantees that at least one tone will be present in each 2-MHz or greater bandwidth normally used for VLBI. Selection of channel center frequencies will assure that one tone will be present in one sideband of each channel for narrower bandwidths.

DSS 25 has a X-band and Ka-band PCG with tone spacing that may be selected as either 4.0 MHz or 1.0 MHz. The tones are derived from the station frequency reference as is the case with the standard phase calibrators.

In the VLBI correlator, a measurement is made of the relative phase delay of the calibration tone in each channel. This is accomplished by correlating the recorded channel with a local (baseband) model of the individual calibration tone. These phase delay measurements enable the phase slope of the entire instrumentation system to be estimated and any phase offsets introduced across the analog RF and IF input to be removed. The result is a group delay calibration with a precision of 0.0015 ns for a 100-s integration time.

In addition to the precise measurement of phase delay in post processing at the VLBI correlator, the phase calibration tones are also monitored in near real-time by the DVP to detect instrument and configuration problems.

2.3.5 Correlation

The JPL VLBI correlator accepts VLBI data in Mark 5 format. The recorded data from two antennas may be cross-correlated. When the recorded data from one antenna are cross-correlated with that of another antenna, an interferometer is formed and the differential delay for that interferometer is determined. These correlated VLBI data are used for several purposes including navigation support, reference frame support, and radio astronomy.

2.4 Experiment Design

The configuration and sequence of events for any wideband VLBI experiment are defined in a VLBI experiment (VEX) file. The VEX file is a format invented by members of the VLBI community to prescribe a complete description of a VLBI experiment, including scheduling, data-taking and correlation. A file for antenna pointing (.nmc) and a file for correlation control (.msf) are also required for each experiment. For more information regarding these interface specifications users may contact their DSN Mission Interface Manager. Parameters from the VEX file are read by the DVP and are translated to the necessary digital signal processing configuration. The antenna pointing file is provided to the Service Preparation Subsystem (SPS). The DVP creates an experiment log file and controls data recording on the Mark 5C disk packs.

The one or two observation bands (L, S, X or Ka) for the experiment are chosen subject to the band availability on the antenna being used. A set of channels is defined for each observation band. In order to achieve high precision, the span of channel frequencies (within the band) should be wide. The two outermost channels determine this precision. The other channels are needed to resolve the phase ambiguity. Simultaneous measurements at S- and X-band enable the differential delay in the ray path to the two observing stations caused by the Earth's ionosphere to be calibrated (Reference 1).

Individual complex channels can be recorded or real channels can be defined in pairs. A lower sideband/upper sideband (LSB/USB) channel pair is created from an individual complex (I/Q) channel. At each complex, 16 complex channels may be used simultaneously resulting in a maximum of 32 real channels. Each of the 16 channel pairs is assigned to one or the other observation band. A typical set of channel-pair frequencies is shown in Table 2 for simultaneous observation in the S and X bands. Also shown in Table 2 is the IF center frequency for complex channel that produces the corresponding channel-pair frequency. It may be noticed that the channel-pair frequency equals the digital baseband downconverter plus that of the first downconverter (2.0 GHz for S band and 8.1 GHz for X band).

Each channel is digitized with a selected depth of 2 bits per sample resolution. The sampling rate is at the nominal Nyquist rate (2 samples per period) for filters that will nominally be 2, 4, 8, or 16 MHz wide. For example, a 16 MHz filter will have a sample rate of

32 Msamples/sec and so with 2 bits/sample it would produce 64 Mbits/sec. The resultant parallel digital streams are formatted, time-tagged and recorded in the Mark 5C (disk array) format.

Table 2. Typical Channels (S and X Bands) for Wideband VLBI

Channel Pair	IF Center Frequency
2217.99 MHz LSB & USB	217.99 MHz
2222.99 MHz LSB & USB	222.99 MHz
2237.99 MHz LSB & USB	237.99 MHz
2267.99 MHz LSB & USB	267.99 MHz
2292.99 MHz LSB & USB	292.99 MHz
2302.99 MHz LSB & USB	302.99 MHz
8210.99 MHz LSB & USB	110.99 MHz
8220.99 MHz LSB & USB	120.99 MHz
8250.99 MHz LSB & USB	150.99 MHz
8310.99 MHz LSB & USB	210.99 MHz
8420.99 MHz LSB & USB	320.99 MHz
8500.99 MHz LSB & USB	400.99 MHz
8550.99 MHz LSB & USB	450.99 MHz
8570.99 MHz LSB & USB	470.99 MHz

Appendix A

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